



B4

INTERNATIONAL APPLICATION PUBLISHED UNDER THE PATENT COOPERATION TREATY (PCT)

(51) International Patent Classification: H04B 1/30, H04B 1/10	A1	(11) International Publication Number: WO 00/11794
		(43) International Publication Date: 02 March 2000 (02.03.2000)

(21) International Application Number:	PCT/EP99/06100	Published
(22) International Filing Date:	18 August 1999 (18.08.1999)	
(30) Priority Data:	9818397.3 25 August 1998 (25.08.1998) GB	
(60) Parent Application or Grant	KONINKLIJKE PHILIPS ELECTRONICS N.V. [/]; O. PHILIPS PATENTVERWALTUNG GMBH [/]; O. MOORE, Paul. A. [/]; O. BARNARD, Michael. E. [/]; O. SCHWEER, Astrid [/]; O. MOODY, Colin, J. ; O.	

- (54) Title: **LOW IF RECEIVER**
 (54) Titre: **RECEPTEUR FI BASSE TENSION**

(57) Abstract

A low IF receiver suitable for use in cellular and cordless telephones and TV tuners, comprises input signal receiving means (10), quadrature related frequency down conversion means including first and second mixing means (14, 15) having first inputs (12, 13) coupled to the input signal receiving means and a local oscillator means (16) for generating a local oscillator signal having a frequency offset from a nominal centre frequency of the input signal coupled to second inputs (18, 19) of the first and second mixing means, a channel selectivity polyphase filtering means (20) for selecting a wanted channel signal from outputs of the first and second mixing means, means (26, 28, 30) for detecting when the quality of the frequency down converted signal becomes unacceptable due to a large adjacent channel interfering component and means responsive to said detection for changing the local oscillator frequency by at least twice the low IF or one channel bandwidth to displace the large interfering component outside the bandwidth of the channel selectivity polyphase filtering means (20) and simultaneously changing the centre frequency of the polyphase filter.

(57) Abrégé

La présente invention concerne un récepteur FI basse tension destiné à être utilisé dans des téléphones cellulaires sans fil et des syntoniseurs de télévision. Ce récepteur comprend un dispositif de réception de signal d'entrée (10), un dispositif de conversion de fréquence par abaissement associé à la quadrature comprenant des premier et second mélangeurs de fréquence (14, 15), présentant des premières entrées (12, 13) couplées au dispositif de réception de signal d'entrée et un oscillateur local (16) produisant un signal d'oscillation local avec un décalage de la fréquence à partir d'une fréquence centrale nominale du signal d'entrée couplé aux secondes entrées (18, 19) des premier et second mélangeurs de fréquence. En outre, ce récepteur FI basse tension comprend également un filtre polyphasé à sélectivité des trajets (20) permettant de sélectionner un signal de voie souhaitée à partir des sorties des premier et second mélangeurs, des dispositifs (26, 28, 30) permettant de détecter si la qualité du signal de conversion de fréquence par abaissement devient inacceptable en raison d'une grande composante à brouillage dans une voie adjacente, ainsi que des dispositifs qui réagissent à cette détection pour modifier la fréquence de l'oscillateur local par au moins le double de la fréquence FI basse tension ou par une bande de fréquence d'un canal pour déplacer la grande composante à brouillage à l'extérieur du filtre polyphasé à sélectivité de trajets (20) et modifier simultanément la fréquence centrale du filtre polyphasé.

BEST AVAILABLE COPY

PCT

WORLD INTELLECTUAL PROPERTY ORGANIZATION
International Bureau



INTERNATIONAL APPLICATION PUBLISHED UNDER THE PATENT COOPERATION TREATY (PCT)

(51) International Patent Classification ⁷ :	A1	(11) International Publication Number: WO 00/11794
H04B 1/30, 1/10		(43) International Publication Date: 2 March 2000 (02.03.00)
(21) International Application Number: PCT/EP99/06100		(81) Designated States: CN, JP, KR, European patent (AT, BE, CH, CY, DE, DK, ES, FI, FR, GB, GR, IE, IT, LU, MC, NL, PT, SE).
(22) International Filing Date: 18 August 1999 (18.08.99)		
(30) Priority Data: 9818397.3 25 August 1998 (25.08.98) GB		Published <i>With international search report.</i>
(71) Applicant: KONINKLIJKE PHILIPS ELECTRONICS N.V. [NL/NL]; Groenewoudseweg 1, NL-5621 BA Eindhoven (NL).		
(71) Applicant (for DE only): PHILIPS PATENTVERWALTUNG GMBH [DE/DE]; Röntgenstrasse 24, D-22335 Hamburg (DE).		
(72) Inventors: MOORE, Paul A.; Prof. Holstlaan 6, NL-5656 AA Eindhoven (NL). BARNARD, Michael E.; Prof. Holstlaan 6, NL-5656 AA Eindhoven (NL). SCHWEER, Astrid; Prof. Holstlaan 6, NL-5656 AA Eindhoven (NL).		
(74) Agent: MOODY, Colin J.; Internationaal Octrooibureau B.V., Prof. Holstlaan 6, NL-5656 AA Eindhoven (NL).		
(54) Title: LOW IF RECEIVER		
(57) Abstract		
<p>A low IF receiver suitable for use in cellular and cordless telephones and TV tuners, comprises input signal receiving means (10), quadrature related frequency down conversion means including first and second mixing means (14, 15) having first inputs (12, 13) coupled to the input signal receiving means and a local oscillator means (16) for generating a local oscillator signal having a frequency offset from a nominal centre frequency of the input signal coupled to second inputs (18, 19) of the first and second mixing means, a channel selectivity polyphase filtering means (20) for selecting a wanted channel signal from outputs of the first and second mixing means, means (26, 28, 30) for detecting when the quality of the frequency down converted signal becomes unacceptable due a large adjacent channel interfering component and means responsive to said detection for changing the local oscillator frequency by at least twice the low IF or one channel bandwidth to displace the large interfering component outside the bandwidth of the channel selectivity polyphase filtering means (20) and simultaneously changing the centre frequency of the polyphase filter.</p>		

FOR THE PURPOSES OF INFORMATION ONLY

Codes used to identify States party to the PCT on the front pages of pamphlets publishing international applications under the PCT.

AL	Albania	ES	Spain	LS	Lesotho	SI	Slovenia
AM	Armenia	FI	Finland	LT	Lithuania	SK	Slovakia
AT	Austria	FR	France	LU	Luxembourg	SN	Senegal
AU	Australia	GA	Gabon	LV	Latvia	SZ	Swaziland
AZ	Azerbaijan	GB	United Kingdom	MC	Monaco	TD	Chad
BA	Bosnia and Herzegovina	GE	Georgia	MD	Republic of Moldova	TG	Togo
BB	Barbados	GH	Ghana	MG	Madagascar	TJ	Tajikistan
BE	Belgium	GN	Guinea	MK	The former Yugoslav Republic of Macedonia	TM	Turkmenistan
BP	Burkina Faso	GR	Greece	MN	Mongolia	TR	Turkey
BG	Bulgaria	HU	Hungary	ML	Mali	TT	Trinidad and Tobago
DJ	Denia	IE	Ireland	MR	Israel	UA	Ukraine
BR	Brazil	IL	Israel	MW	Mauritania	UG	Uganda
BY	Belarus	IS	Iceland	MX	Malawi	US	United States of America
CA	Canada	IT	Italy	NB	Mexico	UZ	Uzbekistan
CF	Central African Republic	JP	Japan	NL	Niger	VN	Viet Nam
CG	Congo	KE	Kenya	NO	Netherlands	YU	Yugoslavia
CH	Switzerland	KG	Kyrgyzstan	NZ	Norway	ZW	Zimbabwe
CI	Côte d'Ivoire	KP	Democratic People's Republic of Korea	PL	New Zealand		
CM	Cameroon	KR	Republic of Korea	PT	Poland		
CN	China	KZ	Kazakhstan	RO	Portugal		
CU	Cuba	LC	Saint Lucia	RU	Romania		
CZ	Czech Republic	LI	Liechtenstein	SD	Russian Federation		
DE	Germany	LK	Sri Lanka	SE	Sudan		
DK	Denmark	LR	Liberia	SG	Sweden		
EE	Estonia				Singapore		

Description

5

10

15

20

25

30

35

40

45

50

55

5

1

DESCRIPTION

LOW IF RECEIVER

10

5 Technical Field

15

The present invention relates to a low IF receiver, which expression is intended to cover a low IF receiver section of a transceiver. Such a receiver is suitable for fabrication as an integrated circuit for use in telecommunications products, such as cellular and cordless telephones, and TV tuners.

10

20

Background Art

25

Low IF receivers are frequently implemented as polyphase or sequence asymmetric receivers. Figure 1 of the accompanying drawings illustrates a typical polyphase receiver. A signal received by an antenna 10 is applied to first 15 inputs 12, 13 of quadrature related mixers 14, 15. A local oscillator 16 generating a local oscillator frequency which is offset typically by half a channel spacing from the nominal carrier frequency of the received signal is applied by way of a ninety degree phase shifter 17 to a second input of the mixer 14 and directly to a second input 19 of the mixer 15. The outputs of the mixers 14, 15 20 are the I and Q channels, respectively. These channels are applied to a polyphase filter 20 functioning as a band pass filter. The outputs of the polyphase filter 20 are applied to respective inputs of a demodulator 22 which provides an output, audio frequency signal on an output terminal 24.

30

35

40

50

From Figure 1 it can be seen that the basic polyphase receiver structure 25 is very similar to a zero IF receiver in that the IF consists of a quadrature pair of channels, the in-phase or I channel and the quadrature-phase or Q channel. However, unlike the zero-IF receiver where the local oscillator is run at the centre frequency of the wanted signal, in the polyphase case the low-IF is achieved by running the receiver local oscillator at typically half a channel 30 spacing from the centre frequency of the wanted signal. Thus for example in GSM (Global System for Mobile Communications) with a channel spacing of 200 kHz, if the desired signal is at 950 MHz then the local oscillator could be at

55

5

950.1 MHz giving an IF of 100 kHz. The principal advantage of the polyphase approach over the zero-IF approach is that large DC offsets which are unavoidably generated by the two mixers are no longer in the IF bandwidth and therefore do not interfere with the demodulation of the wanted signal.

10

5 From the example frequencies given it is obvious that any signal present at 950.2 MHz will form an image signal also at an IF frequency of 100 kHz. In principle this is not a problem since the polyphase filter is able to distinguish signals on its complex inputs having opposing phase relationships, as is the case between the desired and the image frequency. However, this image 10 rejection capability is in practice limited by the level of matching and balance that can be achieved in amplitude and phase between the I and Q channels, which means that an extremely strong signal at the image frequency will cause interference to the desired signal in the IF if the desired signal is too weak. Such a problem does not appear too serious for the DECT and GSM systems 15 20 because there is a relatively relaxed specification on the adjacent channel rejection required by a receiver. However if the polyphase approach is extended to other systems not having such a relaxed specification on the adjacent channel rejection then image rejection will become more of an issue. 25 30

20 Disclosure of Invention

35

An object of the present invention is to improve image rejection in low IF receivers.

40

According to one aspect of the present invention there is provided a low IF receiver comprising frequency down conversion means including a local oscillator, a channel selectivity polyphase filtering means for selecting a wanted channel signal from the output of the frequency down conversion means, means for detecting when the quality of the frequency down converted signal becomes unacceptable and means responsive to said detection for changing the local oscillator frequency and the centre frequency of the polyphase filter.

45

30 More particularly the present invention provides a low IF receiver comprising input signal receiving means, quadrature related frequency down conversion means including first and second mixing means having first inputs 50

55

5

coupled to the input signal receiving means and second inputs coupled to a local oscillator means for generating a local oscillator signal having a frequency offset from a nominal centre frequency of the input signal, a channel selectivity polyphase filtering means for selecting a wanted channel signal from outputs of the first and second mixing means, means for detecting when the quality of the frequency down converted signal becomes unacceptable and means responsive to said detection for changing the local oscillator frequency and the centre frequency of the polyphase filter.

10

5 the first and second mixing means, means for detecting when the quality of the frequency down converted signal becomes unacceptable and means responsive to said detection for changing the local oscillator frequency and the centre frequency of the polyphase filter.

15

In one embodiment of the present invention the means for changing the local oscillator frequency also inverts the sign of one of the local oscillator signals. The inversion of one of the local oscillator signals effectively tunes the polyphase filter by twice the IF.

20

In another embodiment of the present invention the detecting means monitors the potential image frequencies and causes the local oscillator frequency changing means to change the local oscillator frequency in response to deciding which image frequency to use.

25

The detecting means may comprise a wideband polyphase filter, means for identifying the frequencies present in the output of the wideband polyphase filter, and image frequency analysing means for providing a control signal for altering the frequency of the local oscillator and the centre frequency of the channel selectivity polyphase filtering means and the wideband polyphase filter.

30

35 By being able to simultaneously alter the local oscillator frequency and the centre frequency of channel selectivity filter and other complex signal processing means, such as signal de-rotation rate, it is possible to allow the image frequency to be dynamically steered away from that of any unwanted strong signals. This feature can be used to enhance the performance of inexpensive receivers so that they can meet more stringent type approval requirements or alternatively enable high performance products to be built which would have an on-the-air performance better than that which can be met with conventional superheterodyne and homodyne designs.

40

45 According to another aspect of the present invention there is provided an integrated circuit comprising quadrature related frequency down conversion

40

45

50

5

means including first and second mixing means having first inputs for an input signal and second inputs for a local oscillator signal having a frequency offset from a nominal centre frequency of the input signal, a channel selectivity polyphase filtering means for selecting a wanted channel signal from outputs of the first and second mixing means, means for detecting when the quality of the frequency down converted signal becomes unacceptable and means responsive to said detection for providing a control signal for changing the local oscillator frequency and the centre frequency of the polyphase filter.

10

5 the first and second mixing means, means for detecting when the quality of the frequency down converted signal becomes unacceptable and means responsive to said detection for providing a control signal for changing the local oscillator frequency and the centre frequency of the polyphase filter.

15

According to a further aspect of the present invention there is provided a 10 method of operating a low IF receiver, comprising the steps of monitoring the quality of a frequency down-converted signal; if the signal quality is unacceptable, changing the local oscillator frequency by the equivalent of a channel bandwidth; determining if the signal quality is better; and if not, reverting back to the prior local oscillator frequency.

25

15

Brief Description of the Drawings

30

The present invention will now be described, by way of example, with reference to the accompanying drawings, wherein:

Figure 1 is a block schematic diagram of the already described zero IF 20 polyphase receiver.

35

Figure 2 is a block schematic diagram of one embodiment of a low IF receiver made in accordance with the present invention,

40

Figure 3 shows a typical polyphase filter which can be used in the embodiment shown in Figure 2,

Figures 4A and 4B illustrate frequency distribution at RF,

Figures 5A and 5B illustrate infradyne operation,

Figures 6A and 6B illustrate supradyne operation, and

45

Figure 7 is a flow chart of an algorithm which effects a change of frequency conversion mode.

30 In the drawings the same reference numerals have been used to indicate corresponding features.

50

55

5

Modes for Carrying Out the Invention

10

Compared to Figure 1, the embodiment of the invention shown in Figure 2 includes additional features, namely a wideband polyphase filter 26, Fast Fourier Transform (FFT) stage 28 and a microcontroller 30, which enable the

15

5 receiver to monitor potential image frequencies and make an optimal choice at any instant of an IF and the consequent image frequency to use. One method of changing the sign of the IF in one channel, the Q channel, whilst leaving the sign of the other channel, the I channel, unchanged is to invert the phase of the local oscillator 16 output by providing a switchable inverter 34 in the local 10 oscillator path to the input 19 of the mixer 15. An alternative method of changing the sign of the IF in one channel whilst leaving the sign of the other channel, the I channel, unchanged is to locate the inverter 34 in the output path of the mixer 15, as shown in broken lines, thus inverting the Q channel. 15 Optionally the inverter 34 may also be used in the I channel. Digital 20 implementations are also possible. Irrespective of the actual method of changing the sign, this always has the effect of retuning the polyphase filters 20 and 26.

25

30

More particularly the I and Q channels are coupled to the polyphase filter 26 which has a bandwidth wide enough to allow through the IF signals 25 corresponding to potential image frequencies, which might extend for instance to three channels either side of the wanted signal. Since there is only a requirement to detect strong signals in this frequency range this filter 26 does not need a very high dynamic range, rather only enough sensitivity to guarantee 30 detecting reliably an image signal that would cause an image rejection problem.

35

40 This means that the polyphase filter 26 does not require so much resource in terms of silicon area or DSP instruction cycles as does the channel selectivity polyphase filter 20.

45

Figure 3 illustrates an embodiment of a 5th order polyphase filter 36 30 which can be used as filter 20 and/or 26 in Figure 2. The filter comprises 2 sets of stages, corresponding stages in each set being identical. For convenience of description only one set of the stages will be described and the corresponding components in the non-described set will be shown in parenthesis. A first stage 50

55

5

of the filter comprises a source resistor R1 (R3), a capacitor C1 (C6) and a transconductor JC1 (JC9) coupled in parallel between signal rails 38, 40. A cross-coupled gyrator consisting transconductors JC26 and JC17 is coupled to the signal rails 38, 40. A current source J1 (J2) representing the inputs is coupled to the signal rails 38, 40. The current sources J1, J2, correspond to signal inputs from the mixers 14, 15 (Figure 2).

10

5

The second, third and fourth stages are of identical layout and will be described collectively. A transconductor JC2 (JC10), JC4 (JC12), JC6 (JC14) is connected in parallel with a capacitance C2 (C7), C3 (C8), C4 (C9) and another

15

10

transconductor JC3 (JC11), JC5 (JC13), JC7 (JC15) between signal rails 42, 44. Cross coupled gyrators JC19 (JC18), JC21 (JC20) and JC23 (JC22) are

20

15

Cross coupled gyrators JC19 (JC18), JC21 (JC20) and JC23 (JC22) are coupled to the signal rails 42, 44 of the respective stages. The fifth stage of the filter comprises a transconductor JC8 (JC16), capacitance C5 (C10) and resistor

25

20

R2 (R4) connected in parallel between signal lines 46, 48. A cross coupled gyrator JC25 (JC24) is coupled to the signal rails 46, 48. Outputs are derived from the fifth stages. In practice the transconductors JC17 to JC26 are all controlled by a common current control circuit that is adjusted to tune the centre frequency of the filter 36.

30

25

Reverting to Figure 2, the signal from the wideband polyphase filter 26 is

35

30

applied to the FFT stage 28 which is sufficiently refined as to identify which channels potential interfering signals are located in. Finally the microcontroller 30 analyses the signal strengths present in all of the potential image channels and selects the appropriate IF frequency to minimise potential image breakthrough by controlling the local oscillator 16 frequency and tuning the

40

35

polyphase filters 20 and 26 to the appropriate centre frequency using signals on a common control line 32.

45

If there is some latitude in the choice of IF frequency then other considerations, such as the lowest power setting of the receiver, may then also be taken into account in the final IF frequency choice.

50

30

The present invention is based on operating the frequency down conversion means, that is the mixers 14, 15 in two alternative modes, one in which the local oscillator frequency f_{LO} is greater than the nominal RF channel

5

10

frequency f_{Ch} , so called supradyne operation and another in which f_{Lo} is lower than f_{Ch} , so called infradyne operation. In a simple mode of operation the microcontroller 30 switches the local oscillator frequency f_{Lo} by twice the IF, that is by $2f_{IF}$, to identify which channels the potential interfering signals are located in.

15

In order to illustrate the operation of the receiver made in accordance with the present invention reference is made to Figures 4A, 4B, 5A, 5B, 6A and 6B.

20

Figures 4A and 4B show the frequency distribution at RF with the stronger interfering components being below the wanted channel f_{Ch} in Figure 4A and above the wanted channel in Figure 4B. The drawings also indicate the local oscillator frequencies $f_{Lo,I}$ and $f_{Lo,S}$ and their negative counterparts for infradyne and supradyne operation.

25

Figures 5A and 5B show the frequency distributions at IF after quadrature conversion for infradyne operation. The bandwidth of the channel selectivity polyphase filter 20 is referenced B_{20} . The broken line indicates the practically limited rejection on the image side. By comparing Figures 5A and 5B it will be noted that in Figure 5A the larger interfering component lies within the limited image rejection of the filter 20 whereas in Figure 5B it is the smaller interfering component.

35

Figures 6A and 6B show the frequency distributions at IF after quadrature conversion for supradyne operation. In Figure 6A, the smaller interfering component lies within the limited bandwidth B_{20} of the filter 20 whereas in Figure 6B it is the larger component within the limited bandwidth B_{20} of the filter.

40

By making the bandwidth of the wideband polyphase filter 26 equal to at least 3 channel bandwidths it is possible to determine where a larger interfering component lies and to eliminate it by simply selecting a local oscillator frequency f_{Lo} to give infradyne or supradyne operation, and changing the centre frequency of the filter 26 accordingly.

50

Figure 7 is a flow chart of an algorithm which effects a change of frequency conversion mode in the case of poor quality reception.

55

5

Block 50 denotes monitoring signal quality by comparing the signal with a threshold value, for example by considering the bit error rate in a repeatable signal such as a training sequence and/or synchronisation code word, averaged over a predetermined number of bursts or repeats. In block 52 a check is made to see if the measured quality is unacceptable. If the answer is No (N) then the flow chart reverts to the block 50. However if the answer is Yes (Y), then in block 54 the local oscillator frequency is changed by the equivalent of one channel bandwidth, for example 200kHz in the case of GSM. In block 56 a figure of merit is determined and in block 58 a check is made to see if the reception is better. If the answer is Yes (Y) the flow chart reverts to the block 50. If the answer is No (N), in block 60 the local oscillator frequency is changed back to its original value and the flow chart reverts to the block 50.

10

5 to see if the measured quality is unacceptable. If the answer is No (N) then the flow chart reverts to the block 50. However if the answer is Yes (Y), then in block 54 the local oscillator frequency is changed by the equivalent of one channel bandwidth, for example 200kHz in the case of GSM. In block 56 a figure of merit is determined and in block 58 a check is made to see if the 10 reception is better. If the answer is Yes (Y) the flow chart reverts to the block 50. If the answer is No (N), in block 60 the local oscillator frequency is changed back to its original value and the flow chart reverts to the block 50.

15

15 to see if the measured quality is unacceptable. If the answer is No (N) then the flow chart reverts to the block 50. However if the answer is Yes (Y), then in block 54 the local oscillator frequency is changed by the equivalent of one channel bandwidth, for example 200kHz in the case of GSM. In block 56 a figure of merit is determined and in block 58 a check is made to see if the 10 reception is better. If the answer is Yes (Y) the flow chart reverts to the block 50. If the answer is No (N), in block 60 the local oscillator frequency is changed back to its original value and the flow chart reverts to the block 50.

20

In operation the figure of merit of reception should be verified periodically because accumulating changes in operation may lead to the threshold being too small. Using the quality of the previously received signal, the threshold could be increased for a predetermined period.

25

30 In situations of failing reception leading to the loss of a call-in-progress and a failed connection, the mode of frequency conversion should be changed.

35

In the event of finding big interferers before and after shifting the IF by 20 twice the IF frequency and the microcontroller 30 not being able to decide on the better IF, it can as a final option operate the receiver as a zero IF receiver by 35 shifting the IF by one IF and altering the polyphase filter characteristic so that it appears effectively as 2 filters. This may be done by adjusting the cross coupling, that is, between transconductors JC17 and JC26; JC18 and JC19, 40 JC20 and JC21, JC22 and JC23, and JC24 and JC25 so that the currents are zero.

40

45 In the present specification and claims the word "a" or "an" preceding an element does not exclude the presence of a plurality of such elements. Further, the word "comprising" does not exclude the presence of other elements or steps 30 than those listed.

45

50 From reading the present disclosure, other modifications will be apparent to persons skilled in the art. Such modifications may involve other features

55

5

which are already known in the design, manufacture and use of low IF receivers and component parts thereof and which may be used instead of or in addition to features already described herein.

10

5 Industrial Applicability

Low IF radio receivers and integrated circuits comprising such receivers.

15

20

25

30

35

40

45

50

55

Claims

5

10

15

20

25

30

35

40

45

50

55

5

CLAIMS

- 10 1. A low IF receiver comprising frequency down conversion means including a local oscillator, a channel selectivity polyphase filtering means for selecting a wanted channel signal from the output of the frequency down conversion means, means for detecting when the quality of the frequency down converted signal becomes unacceptable and means responsive to said detection for changing the local oscillator frequency and the centre frequency of the polyphase filter.
- 15 10 2. A low IF receiver comprising input signal receiving means, quadrature related frequency down conversion means including first and second mixing means having first inputs coupled to the input signal receiving means and second inputs coupled to a local oscillator means for generating a local oscillator signal having a frequency offset from a nominal centre frequency of the input signal, a channel selectivity polyphase filtering means for selecting a wanted channel signal from outputs of the first and second mixing means, means for detecting when the quality of the frequency down converted signal becomes unacceptable and means responsive to said detection for changing the local oscillator frequency and the centre frequency of the polyphase filter.
- 20 25 30 35 3. A receiver as claimed in claim 2, characterised in that said means changes the local oscillator frequency by at least twice the low IF.
- 40 45 4. A receiver as claimed in claim 3, characterised in that the means for changing the local oscillator frequency also inverts the sign of the local oscillator signal.
- 50 5. A receiver as claimed in claim 3, characterised in that said means in response to finding that an unacceptable signal is present at both the respective local oscillator frequencies, adapts the polyphase filtering means to operate as two zero IF filters.

5

10

6. A receiver as claimed in claim 2, characterised in that the detecting means monitors image frequencies and causes the local oscillator frequency changing means to change the local oscillator frequency in response to deciding which image frequency to use.

15

20

7. A receiver as claimed in claim 1, characterised in that the detecting means comprises a wideband polyphase filter, means for identifying the frequencies present in the output of the wideband polyphase filter, and image frequency analysing means for providing a control signal for altering the frequency of the local oscillator and the centre frequency of the channel selectivity polyphase filtering means and the wideband polyphase filter.

25

30

8. A receiver as claimed in claim 2, characterised in that the detecting means comprises a wideband polyphase filter, means for identifying the frequencies present in the output of the wideband polyphase filter, and image frequency analysing means for providing a control signal for altering the frequency of the local oscillator and the centre frequency of the channel selectivity polyphase filtering means and the wideband polyphase filter.

20

35

40

45

9. An integrated circuit comprising quadrature related frequency down conversion means including first and second mixing means having first inputs for an input signal and second inputs for a local oscillator signal having a frequency offset from a nominal centre frequency of the input signal, a channel selectivity polyphase filtering means for selecting a wanted channel signal from outputs of the first and second mixing means, means for detecting when the quality of the frequency down converted signal becomes unacceptable and means responsive to said detection for providing a control signal for changing the local oscillator frequency and the centre frequency of the polyphase filter.

30

50

10. A method of operating a low IF receiver, comprising the steps of: monitoring the quality of a frequency down-converted signal;

55

5

if the signal quality is unacceptable, changing the local oscillator frequency by the equivalent of a channel bandwidth;

10

determining if the signal quality is better; and

if not, reverting back to the prior local oscillator frequency.

15

20

25

30

35

40

45

50

55

1/4

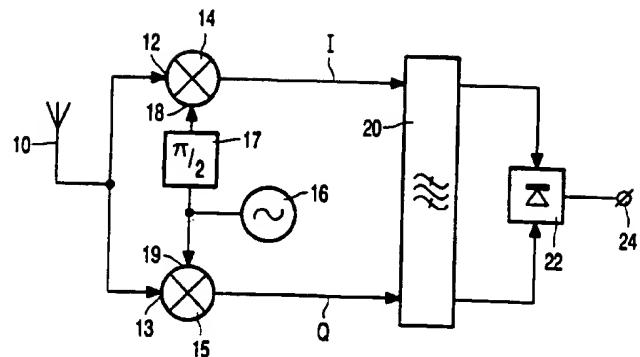


FIG. 1

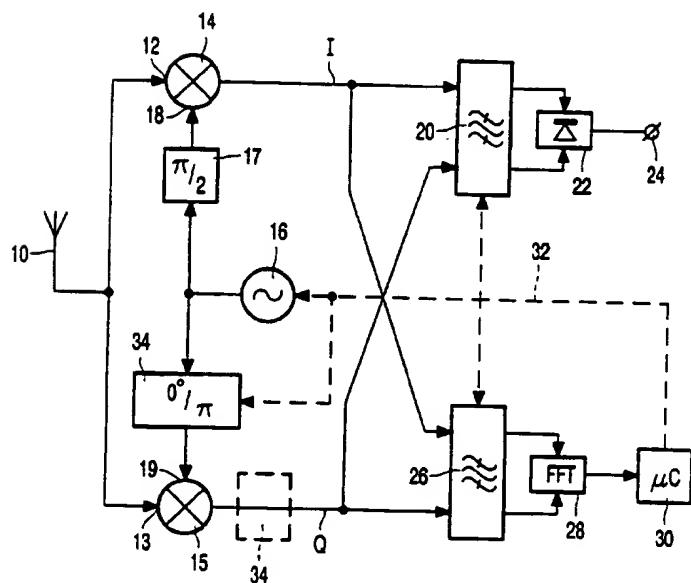


FIG. 2

2/4

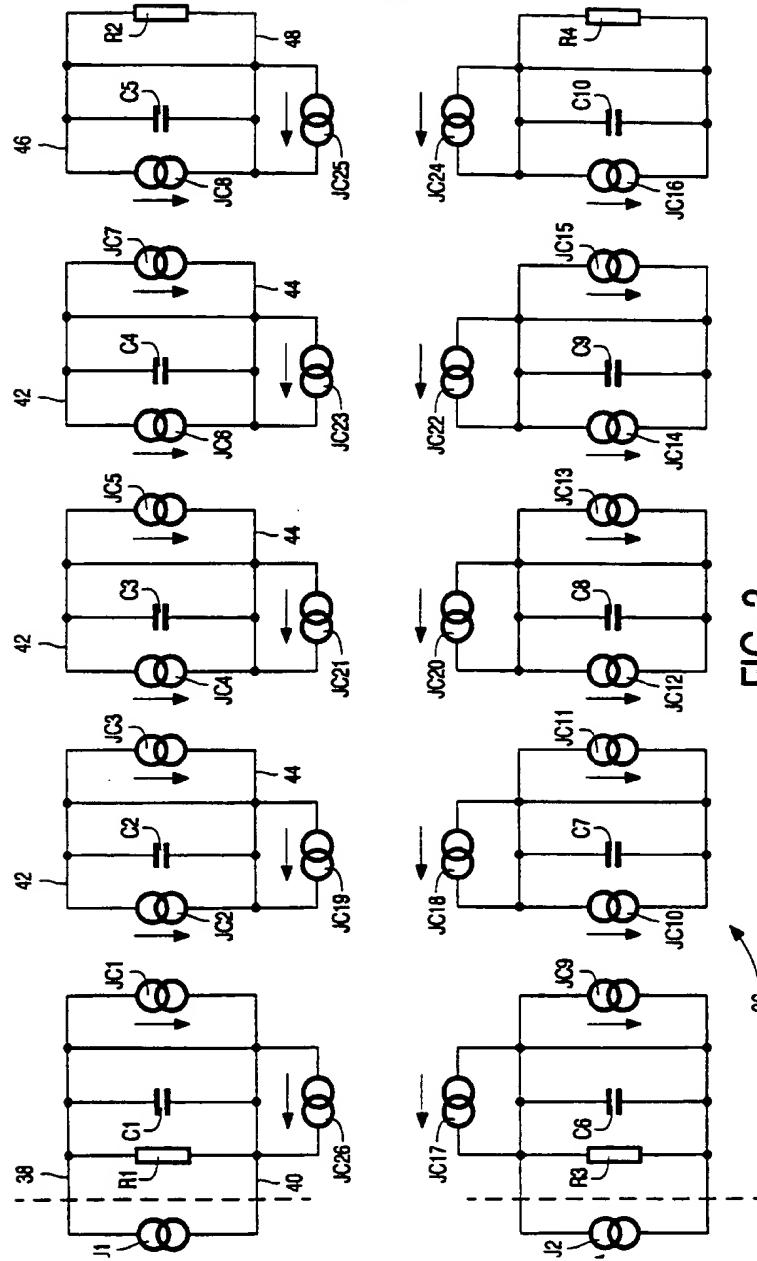


FIG. 3

3/4

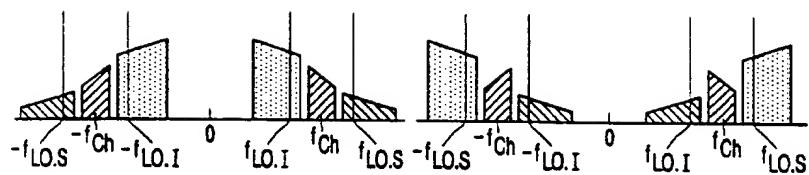


FIG. 4A

FIG. 4B

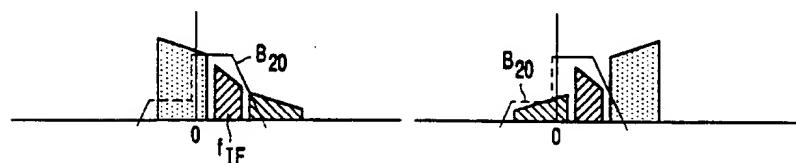


FIG. 5A

FIG. 5B



FIG. 6A

FIG. 6B

4/4

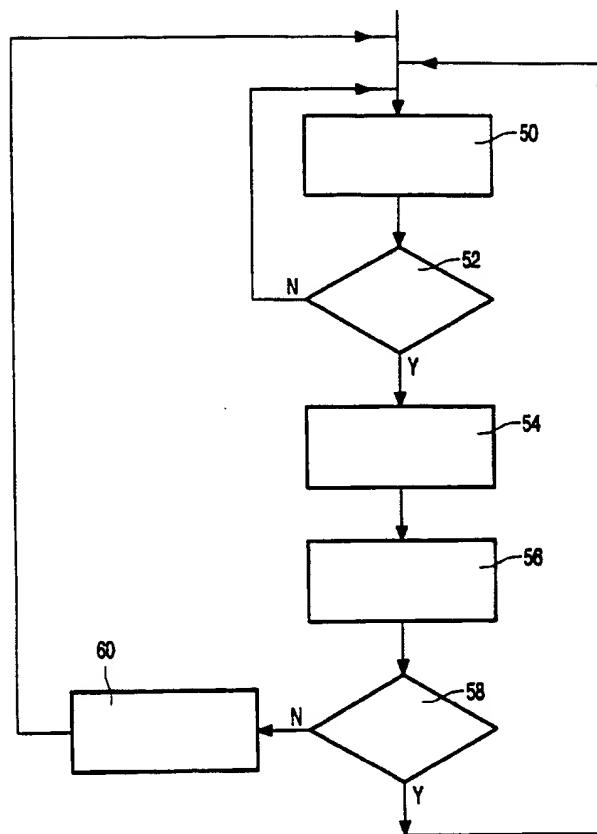


FIG. 7

INTERNATIONAL SEARCH REPORT

Int'l. Appl. No.
PCT/EP 99/06100

A. CLASSIFICATION OF SUBJECT MATTER
IPC 7 H04B1/30 H04B1/10

According to International Patent Classification (IPC) or to both national classification and IPC

B. FIELDS SEARCHED

Minimum documentation searched (classification system followed by classification symbols)
IPC 7 H04B

Documentation searched other than minimum documentation to the extent that such documents are included in the fields searched

Electronic data base consulted during the international search (name of data base and, where practical, search terms used)

C. DOCUMENTS CONSIDERED TO BE RELEVANT

Category *	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
A	DE 43 03 903 A (HEINZMANN GUSTAV DR ING) 24 February 1994 (1994-02-24) abstract; figures 1-4 ----	1-10
A	EP 0 797 292 A (PHILIPS ELECTRONICS NV) 24 September 1997 (1997-09-24) column 3, line 27 -column 5, line 6; figure 3 ----	1-10
A	EP 0 542 520 A (NOKIA MOBILE PHONES LTD) 19 May 1993 (1993-05-19) column 2, line 34 -column 3, line 17; figures 3,4 -----	1-4, 9, 10

Further documents are listed in the continuation of box C.

Patent family members are listed in annex.

* Special categories of cited documents:

- "A" document defining the general state of the art which is not considered to be of particular relevance
- "E" earlier document but published on or after the international filing date
- "L" document which may throw doubt on novelty claim(s) or which is cited to establish the publication date of another citation or other special reason (as specified)
- "O" document referring to an oral disclosure, use, exhibition or other means
- "P" document published prior to the international filing date but later than the priority date claimed

- "T" later document published after the international filing date or priority date and not in conflict with the application but cited to understand the principle or theory underlying the invention
- "X" document of particular relevance; the claimed invention cannot be considered novel or cannot be considered to involve an inventive step when the document is taken alone
- "V" document of particular relevance; the claimed invention cannot be considered to involve an inventive step when the document is combined with one or more other such documents, such combination being obvious to a person skilled in the art
- "S" document member of the same patent family

Date of the actual completion of the international search

17 November 1999

Date of mailing of the international search report

25/11/1999

Name and mailing address of the ISA
European Patent Office, P.B. 5818 Patentlaan 2
NL - 2280 HV Rijswijk
Tel. (+31-70) 340-2040, Tx. 31 651 ecp nl.
Fax: (+31-70) 340-3016

Authorized officer

Andersen, J.G.

INTERNATIONAL SEARCH REPORT

Information on patent family members

				International Application No	
				PCT/EP 99/06100	
Patent document cited in search report	Publication date	Patent family member(s)		Publication date	
DE 4303903	A 24-02-1994	NONE			
EP 0797292	A 24-09-1997	CN JP	1164782 A 10070482 A	12-11-1997 10-03-1998	
EP 0542520	A 19-05-1993	FI DE DE US	915388 A 69218551 D 69218551 T 5396657 A	15-05-1993 30-04-1997 11-09-1997 07-03-1995	

**This Page is Inserted by IFW Indexing and Scanning
Operations and is not part of the Official Record**

BEST AVAILABLE IMAGES

Defective images within this document are accurate representations of the original documents submitted by the applicant.

Defects in the images include but are not limited to the items checked:

- BLACK BORDERS**
- IMAGE CUT OFF AT TOP, BOTTOM OR SIDES**
- FADED TEXT OR DRAWING**
- BLURRED OR ILLEGIBLE TEXT OR DRAWING**
- SKEWED/SLANTED IMAGES**
- COLOR OR BLACK AND WHITE PHOTOGRAPHS**
- GRAY SCALE DOCUMENTS**
- LINES OR MARKS ON ORIGINAL DOCUMENT**
- REFERENCE(S) OR EXHIBIT(S) SUBMITTED ARE POOR QUALITY**
- OTHER:** _____

IMAGES ARE BEST AVAILABLE COPY.

As rescanning these documents will not correct the image problems checked, please do not report these problems to the IFW Image Problem Mailbox.